

doi:10.15199/48.2023.01.21

Comparative study of Sliding Mode and incremental conductance for Maximum Power Point Tracker for Photovoltaic Array

Abstract. The aim of this paper is to compare two techniques the sliding mode control and incremental conductance, for maximum power point tracking (MPPT) of PV stand-alone system under a rapid change of irradiation and temperature. The comparison study is based on convergence speed, steady-state oscillations, and tracking efficiency. This system consists of a DC/DC boost converter, photovoltaic arrays, load, and MPPT control. The system performance of sliding mode control was compared to the incremental conductance algorithm using the Sim-Power System of MATLAB. From the simulation results the sliding mode method shows a better performance and also has a lower oscillation.

Streszczenie. Celem tego artykułu jest porównanie dwóch technik, sterowania trybem ślizgowym i przewodności przyrostowej, dla śledzenia punktu maksymalnej mocy (MPPT) w autonomicznym systemie PV przy gwałtownych zmianach napromieniowania i temperatury. Badanie porównawcze opiera się na prędkości zbieżności, oscylacjach stanu ustalonego i wydajności śledzenia. System ten składa się z konwertera doładowania DC/DC, paneli fotowoltaicznych, obciążenia i sterowania MPPT i będzie symulowany przy użyciu systemu Sim-Power firmy MATLAB. Techniki te mają na celu monitorowanie parametrów wyjściowych systemu PV i uzyskanie optymalnego cyklu pracy. Niniejsze opracowanie zawiera szczegółową analizę i porównanie różnych technik. (Badanie porównawcze trybu przesuwania i przewodnictwa przyrostowego dla modułu śledzenia punktu maksymalnej mocy dla macierzy fotowoltaicznej)

Keywords: stand-alone, Converter, MPPT, sliding mode, incremental conductance.

Słowa kluczowe: samodzielny, konwerter, MPPT, tryb przesuwany, przewodność przyrostowa.

Introduction

In recent years, the spread of renewable energy resources worldwide are becoming very extensive, which allow people exploiting energy solar to meet their needs on one side and protect the environment on the other side [1]. Conversion efficiency reach of PV arrays from 13 % to 47%, that's not sufficient to meet the needs of consumers[2]. Hence, they increase the number of PV arrays, that affects negatively cost. It becomes necessary to extract as much power as possible from PV arrays, with considering efficacy, cost, and reliability.

This paper presents a comparative study of a DC/DC converter controller in different input information. Many researchers have done for MPPT algorithms, utilizing different control techniques (classic and modern), namely, Perturb and Observe (P&O) [3], incremental conductance [4], hill-climbing[5], Grey wolf optimization [6], genetic algorithm [7], neural networks (ANN) [8], particle swarm optimization (PSO)[9][10], fuzzy logic (FL) [11], predictive control [12], sliding mode (SM) [13], etc. In this current document, we analyze the simulation of photovoltaic system a stand-alone.

The objective of this work is tracking MPP and optimisation the duty cycle of proposed SMC controller, and the comparison study is for ensure the performance of SMC control.

The description of the comparative study is as follows: Description of system is explained in Section 2. In Section 3, describes the important of MPPT techniques. Section 4, describes comparison and analyses the simulation of respective techniques. The conclusion will be provided in Section 5.

DDescription of a stand-alone system

The term "stand-alone PV system" is used to describe a PV system that is not connected to the grid. A PV panel, DC/DC boost converter, load, and a control circuit MPPT are used to operate the system [8][9]. As show in Figure.1:

The description of study PV systeme parameters:

The modules are connected to PV arrays (2 series and 3 parallel). The specifications of the PV module are as detailed belo:

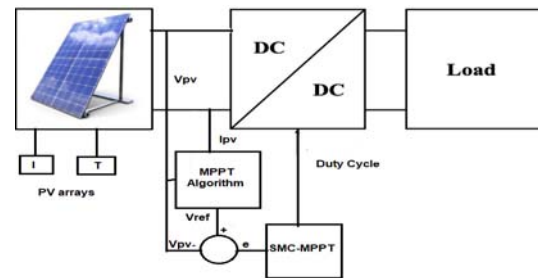


Fig.1. The basic structure photovoltaic system

Table 1. KC200GT Data Sheet under STC Conditions and internal parameters

Number of cells in series of each panel (N_s)	54
Power maximal (P_{max})	200.143 W
Current of short-circuit (I_{sc})	8.21 A
Voltage of circuit open (V_{oc})	32.9 V
Current of MPP (I_{mpp})	7.61 A
Voltage of MPP (V_{mpp})	26.3 V
Resistance series (R_s)	0.0036 Ω
Factor ideal (n)	1.4099
Courant inverse de saturation (I_o)	4.0992 e-7 A
Photo current (I_{ph})	8.2100 A

The DC-DC boost converter is installed between PV arrays and load. The values of the parameters of converter as follows:

Resistance (R_n) = 1 Ω , Inductance (L) = 400 μ H, Capacitance (C_{out}) = 1200 μ F, Capacitance (C_{in}) = 4700 μ F, Diode: Resistance (R_{on}) = 0.001 Ω , Direct voltage (V_d) = 0.8 V, damping resistance (R_a) = 500 k Ω , Mosfet: FET Resistance (R_{on}) = 0.1 Ω , Internal diode resistance (R_d) = 0.01 Ω .

The last component in each electrical circuit is the load, we used Resistance R=350 Ω .

PV Cell

Solar cells are used to convert solar energy directly into electrical power. The PV module is made a group of cells which can be connected on series or parallel. The photovoltaic cells analogous circuit is explain in [13, 14]. A

current generator and a diode connect to series and parallel resistors. The equivalent of a single non-ideal solar cell is shown Figure 2. The mathematical equations (1 to 6) define the output current of a single cell.

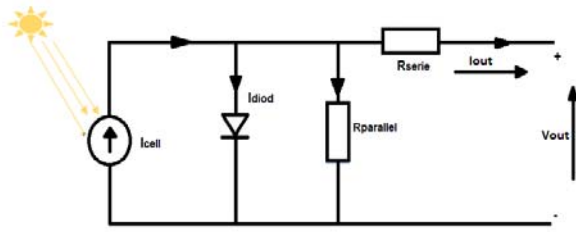


Fig.2. The equivalent circuit model of PV cell.

$$(1) \quad I_{cell} = I_{out} - I_{diod} - I_{parallel}$$

$$(2) \quad I_{diod} = I_0 \left[e^{\left(\frac{V_{out} + I_{out} \cdot R_{serie}}{V_{th}} \right)} - 1 \right]$$

$$(3) \quad I_{out} = I_{Cell} - I_0 \left[e^{\left(\frac{V_{out} + I_{out} \cdot R_{serie}}{V_{th}} \right)} - 1 \right] - \frac{V_{out} + I_{out} \cdot R_{serie}}{R_{parallel}}$$

$$(4) \quad V_{diod} = V_{out} + I_{out} \cdot R_{serie}$$

$$(5) \quad V_{th} = \frac{k \cdot T}{q}$$

$$(6) \quad V_{parallel} = \frac{V_{out} + I_{out} \cdot R_{serie}}{R_{parallel}}$$

I_{out} and V_{out} are cell output current and voltage; I_0 is the cell reverse saturation current; I_{cell} is the light-generated current. R_{serie} and $R_{parallel}$ are series and shunt resistances; q is electronic charge; k is Boltzmann's constant; T is cell temperature in K.

DC/DC boost converter

The boost converter is also defined as the step-up converter [15]. It is generally used in the conversion of a low input voltage to a high output voltage [16]. It consists of DC input voltage source V_{PV} , an inductor L , a switch K , a diode, and two capacitors C_{in} and C_{out} . Figure 7 shows the electrical diagram of the proposed boost converter.

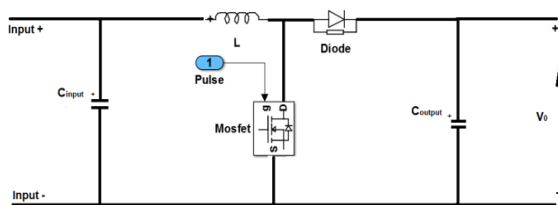


Fig.3. The electrical diagram of the boost converter

The relationship between the input voltage V_{PV} and output V_{out} of the converter is given by the equation (7):

$$(7) \quad V_{out} = \frac{V_{PV}}{D - 1}$$

Equations (8) and (9) respectively give the expression of the capacitance and the inductance filter:

$$(8) \quad C = \frac{D \cdot V_a}{f_s \cdot R \cdot \Delta V}$$

$$(9) \quad L = \frac{D \cdot V_a}{f_s \cdot \Delta I}$$

In these relations, D is cycle rapport, f_s is frequency of the Mosfet, ΔV is voltage limit tolerable, ΔI is tolerable current limit and R is load resistance.

In this simulation, we use $f_s = 1000$ kHz, $\Delta V / V = 5\%$ and $\Delta I / I = 30\%$. The voltage converter offered has two capacities, C_{in} and C_{out} , which have the respective role of reducing the ripple of input and output. For this study, the Mosfet frequency is 1 kHz.

The load

The load is considered constant resistance ($R = 350$ ohms).

MPPT of a stand-alone system (SAS)

Incremental conductance (IC) method

This technique is created to beat the control of P & O MPPT techniques [17]. The IC can determine the MPPT reached and stop perturbing the operating point.

The instantaneous and Incremental conductance of PV arrays utilized to track the MPP at which the maximum power can be tied below as an equation;

$$(10) \quad \frac{I}{V} = \frac{dI}{dV}$$

where dI/dV is the incremental conductance and I/V is the instantaneous conductance.

To get the full control point, must be the power shift and the voltage level ratio are going to zero, according to the derivative rule,

$$(11) \quad \frac{dP}{dV} = \frac{d(V \cdot I)}{dV} = V \frac{dI}{dV} + I$$

At MPP " $dP/dV=0$ " therefore, the instantaneous and incremental conductance becomes opposite and equal as given in (10).

If $dP/dV=0$ (PV voltage is equal to the MPP) duty cycle should be decreased. If $dP/dV < 0$ (PV MPP is less than the voltage) duty cycle must be decreased. And if $dP/dV > 0$, this means the MPP voltage is higher than the PV voltage, and the duty cycle must be augmented [18].

The actual peak power from the PV arrays is determined by IC and P & O. The IC technique is more accurate than P&O technique. In addition, the IC Method shows the exact value of atmospheric change [17].

SMC MPPT technique

It is the theory of structure systems variable (SSV) that allows the task to be accomplished. This control gives stability in results, even with large variations in power or load, and good dynamic, which appeared in the Soviet Union in the 1950 [19][20].

The derivative of power is equal to zero as shown in equation (20) [21]. In the present document, the sliding surface is taken based on equation (12):

$$(12) \quad S = dP$$

The relation expressed the control signal as follows:

$$(13) \quad D = D_{eq} + D_n$$

where U_n is the switching control and U_{eq} is the equivalent control. U_{eq} found from the steady condition and given as below:

If $U=U_{eq} \Leftrightarrow \dot{S}=0$ & $\dot{S}=0$ the derivative of the sliding surface is:

$$(14) \quad \dot{S} = \frac{d}{dt}[dP] = \frac{d}{dt}[d(V.I)]$$

$$(15) \quad \dot{S} = dV \cdot \frac{dI}{dt} + dI \cdot \frac{dV}{dt} = k_1 \cdot \frac{dI}{dt} + k_2 \cdot \frac{dV}{dt}$$

$$(16) \quad k_1 = dV \text{ \& } k_2 = dI$$

The laws of voltage and current in the circuit in two cases (switch off & on) of Figure 3, by applying Kirchoff's function:

$$(17) \quad \frac{dI}{dt} = \frac{V_{pv}}{L} - \frac{V_{out}}{L}(1-D)$$

$$(18) \quad \frac{dV_{out}}{dt} = \frac{I_L}{C}(1-D)$$

Substituting Eq. (17) & (18) into Eq. (15) results as:

$$(19) \quad \dot{S} = k_1 \left[\frac{V_{pv}}{L} - \frac{V_{out}}{L}(1-D) \right] + k_2 \left[\frac{I_L}{C}(1-D) \right]$$

By imposing this is conditions $S=0$ & $\dot{S}=0$ we get:

$$(20) \quad \dot{S}=0 \rightarrow k_1 \left[\frac{V_{pv}}{L} - \frac{V_{out}}{L}(1-D) \right] + k_2 \left[\frac{I_L}{C}(1-D) \right] = 0$$

According to Eq. (7), the control input is taken as:

$$(21) \quad D_{eq(t)} = 1 - \frac{V_{pv}}{V_{out}}$$

The trajectory of the sliding surface is so essential, that the U_n is selected for the stability criterion ($\dot{S}.S < 0$).

$$(22) \quad D_{n(t)} = \frac{V_{pv}}{V_{out}} - B$$

Where B is calculated according to Lyapunov's stability criteria and considered as a control signal.

Substituting Eq. (21) and Eq. (22) into Eq. (23) as follows:

$$(23) \quad D_t = 1 - B$$

The Lyapunov function adopted to explain the stability

$$(24) \quad P = \frac{1}{2} S^2 \text{ Then : } \dot{P} = S \dot{S}$$

Replace Eq. (19) and Eq. (24) with Eq. (25), which shows the following result:

$$(25) \quad \dot{P} = \left(k_1 \left[\frac{V_{pv}}{L} + \frac{V_{out}}{L} B \right] - k_2 \left[\frac{I_L}{C} B \right] \right) S$$

$$(26) \quad \dot{P} = \left(B \left[\frac{V_{out}}{L} k_1 - \frac{I_L}{C} k_2 \right] + \frac{V_{pv}}{L} k_1 \right) S$$

The convergence result is given below.

$$\left\{ \begin{array}{l} \text{if } S < 0 \Rightarrow B < \frac{V_{pv} \cdot k_1 \cdot C}{V_{out} \cdot k_1 \cdot C - I_L \cdot k_2 \cdot L} \\ \text{if } S > 0 \Rightarrow B > \frac{V_{pv} \cdot k_1 \cdot C}{V_{out} \cdot k_1 \cdot C - I_L \cdot k_2 \cdot L} \\ \text{if } S = 0 \Rightarrow B = \frac{V_{pv} \cdot k_1 \cdot C}{V_{out} \cdot k_1 \cdot C - I_L \cdot k_2 \cdot L} \end{array} \right.$$

Simulation Results and Discussion

The simulation is implementing used three PV module KC200GT in series and tow parallel connection, which are monitored by INC and SMC controllers. The controllers are monitoring the performance of the stand-alone system which has been designed and simulated in Matlab / Sim Power.

The irradiation and the temperature used are variable as shown in figures (4&5) and the load is considered constant $R = 350\Omega$.

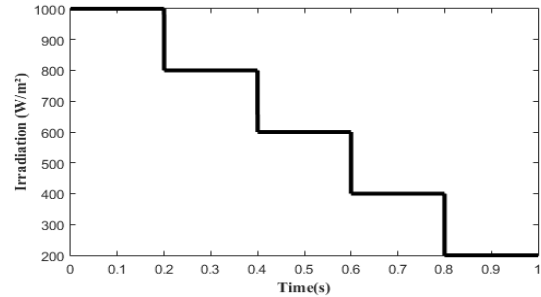


Fig.4. The variation of irradiation

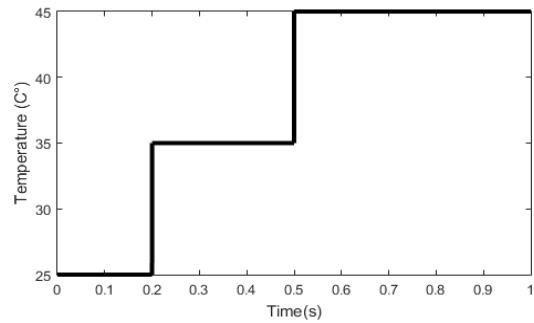


Fig.5. Variation of Temperature

The irradiations were varies as chosen in figure (4) and the temperature was set at $T=25^\circ$, where the curves show the results in Figures (6&7):

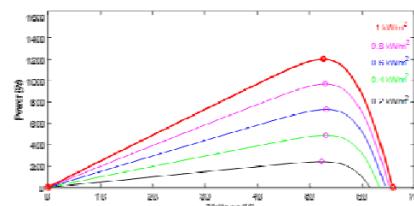


Fig.6. The Voltage-Power characteristic with different irradiation

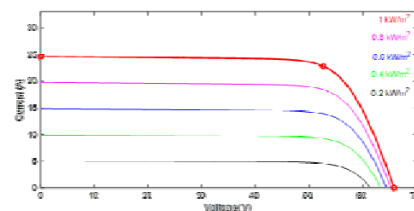


Fig.7. The Current-Voltage characteristic with different irradianations

The temperatures varied as chosen in figure (5) and the irradiation was set at 1000 W/m^2 , where the curves show the results in Figures (8&9):

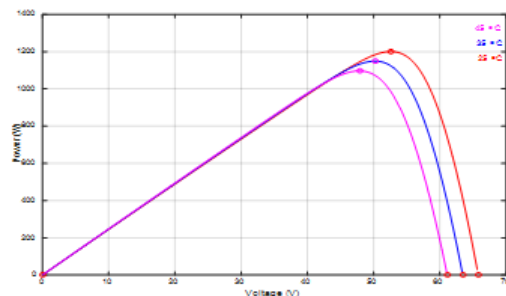


Fig.8. The Voltage-Power characteristic with different temperatures

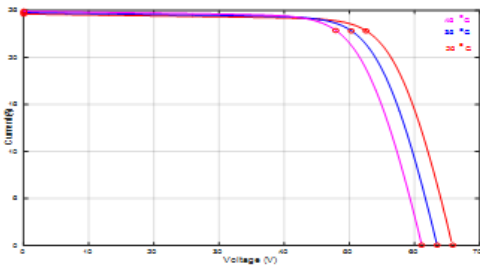


Fig.9. The Voltage-Current characteristic with different temperatures

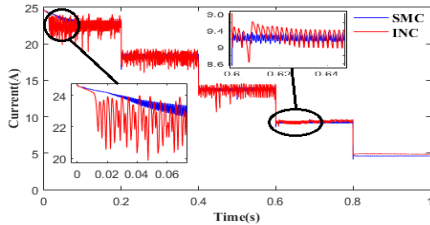


Fig.10. Output Current of PV KC200GT arrays

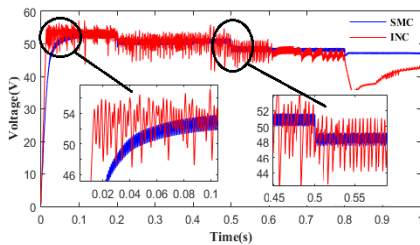


Fig.11. The output voltage of PV KC200GT arrays

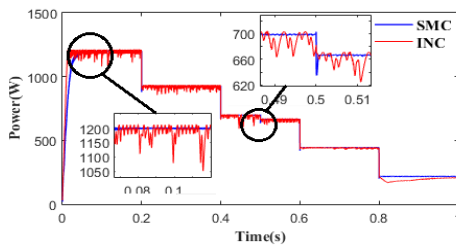


Fig.12. Power of PV KC200GT arrays

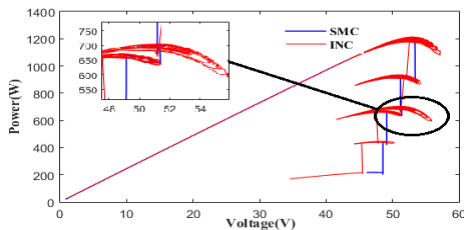


Fig.13. Dynamic MPPT of PV KC200GT arrays

As changing quickly of atmospheric conditions, the simulation results obtained on the MPP for the sliding mode control and INC. The sliding mode control (SMC) give good response in Figures (11) and show the ability to stabilize the voltage in adjacent the voltage required. The INC is reaching the required voltage with chattering and after 0.8 second, the voltage drops below requirement.

Figure (10) shows the output current when the temperature increase and the irradiance decrease automatically. The current drops in two controllers. The SMC gives a good response and without high perturbation than INC controller.

The objective of this paper is obtaining the maximum power from PV arrays. In Figure (12), SMC controller performs accurate in tracking and stability when the is a changing in climatic condition. While INC controller have a high perturbation during climatic changing and around MPP. Also INC controller can't keep the requirement of load during low irradiation or high temperature condition.

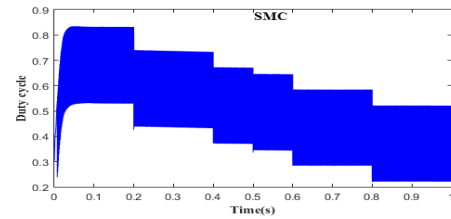


Fig.14. The duty cycle of PV arrays by sliding mode control

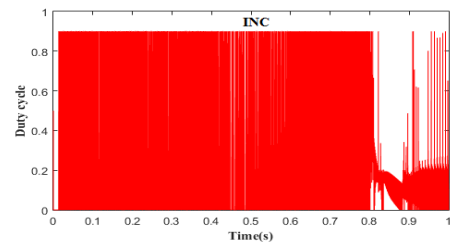


Fig.15. The duty cycle of PV by incremental conductance

The deferent is so clearly in figures (14, 15), the optimum duty cycle of SMC shows the same reaction on each step, and the trucking graph proposed in figure (4), a robust and fast response.

The duty cycle of INC has a perturbation not stable, not reliable and loss the control after 0.8 s.

Conclusion

According to the results of the simulation ,the sliding mode control provides a better design in stand-alone PV arrays system. And its performs significantly better under high variation of environmental condition comparative to conventional INC-MPPT, in terms of overshoot, ripple, speed and accuracy. Moreover, the SMC-MPPT effectively enhances the dynamic responsiveness. The precise tracking of the proposed technique is also shown in the simulation.

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